Chapter 2 Problems

(2-1) Use the conversion formula $\overline{V} = V \angle \phi$

(a)
$$\overline{V_1} = 100 \angle -30^{\circ} \text{ V}$$

(b)
$$\overline{V_2} = 100 \angle 30^{\circ} \text{ V}$$

(2-2) We have,

$$Ri(t) + L\frac{di(t)}{dt} + \frac{1}{C}\int i(t)dt = \sqrt{2}V\cos(\omega t).$$

Taking Laplace Transform we get,

$$RI(s) + LsI(s) + \frac{1}{Cs}I(s) = \sqrt{2}V \times \frac{s}{s^2 + \omega^2}$$
$$I(s) \left[R + Ls + \frac{1}{Cs} \right] = \sqrt{2}V \times \frac{s}{s^2 + \omega^2}$$

$$I(s) = \frac{\sqrt{2}V \times Cs^{2}}{(s^{2}LC + RCs + 1)(s^{2} + \omega^{2})}$$
$$= \frac{\sqrt{2}\frac{V}{L}s^{2}}{(s^{2} + (\frac{R}{L})s + \frac{1}{LC})(s^{2} + \omega^{2})}$$

Solving the above form by partial fractions and then taking the inverse Laplace transform we get

$$i(t) = 0.0368cos(\omega t) + 0.466sin(\omega t) = 0.468cos(\omega t - 85.48)A$$

$$(2-3)$$

$$Z_{L} = |Z_{L}| \angle \phi$$

$$|Z_{L}| = \sqrt{R^{2} + \left(\omega L - \frac{1}{\omega C}\right)^{2}}$$

$$\phi = \tan^{-1} \left[\frac{\omega L - \frac{1}{\omega C}}{R}\right]$$

$$= 85.48^{\circ}$$

$$\bar{I} = \frac{\bar{V}}{|Z_{L}|} \angle - \phi$$

$$= \frac{120 \angle 0}{1.5^{2} + \left(377 \times 20 \times 10^{-3} - \frac{1}{377 \times 100 \times 10^{-6}}\right)^{2}} \angle - 85.48^{\circ}$$

$$= 0.33\angle - 85.48^{\circ}$$

$$\therefore i(t) = 0.33\sqrt{2}cos(\omega t - 85.48)A$$

$$= 0.467cos(\omega t - 85.48) A$$

(2-4)

$$\overline{V} = 90 \angle 30^{\circ}$$

$$\overline{I} = 5 \angle 15^{\circ}$$
 If $\overline{V}_{new} = 100 \angle 0^{\circ}$

$$\overline{I}_{new} = (\frac{\overline{V}_{new}}{\overline{V}})\overline{I} = 5.56 \angle - 15^{\circ}$$

$$(2-5)$$

$$\overline{V} = 100 \angle 0^{\circ} V$$

$$Z = j0.1 + \frac{(2)(-j5)}{2 - j5}$$

$$= 1.82 \angle - 18.9^{\circ} A$$

$$\overline{I} = \frac{100 \angle 0^{\circ}}{1.82 \angle - 18.9^{\circ}}$$

$$= 54.95 \angle 18.9^{\circ} A$$

$$P = VI\cos\phi = 5198.74W$$

$$Q = VI\sin\phi = 1779.93VAR$$

As the current leads the voltage ,so the net load impedance is capacitive hence it draws negative reactive power So net Q=-1779.93VAR.

$$p.f. = \cos\phi = 0.946(Leading)$$

Let $X_1 = 0.1$ and $X_2 = 5$. Then,

$$Q_{X_1} = I^2 X_1 = 301.95 VAR$$

 $Q_{X_2} = I_2^2 X_2$

where I_2 is the current through the capacitance.

$$\overline{I_2} = \overline{I} \times \frac{2}{2 - j5} = 20.41 \angle 87.1^{\circ} A$$

$$Q_{X_2} = 5 \times (20.41)^2 = 2082.84 VAR$$

We further find that,

$$Q_{X_1} + (-Q_{X_2}) = -1780.57VAR \approx Q$$

 Q_{X_2} takes a negative sign as the capacitor supplies reactive power.

$$\therefore Q = \sum_{k} I_{k}^{2} X_{k}$$

(2-6)

$$|I_s| = \frac{V}{\sqrt{R_s^2 + X_s^2}}$$

For the series circuit, the real power and the reactive power supplied to the load are

$$P_s = |I_s|^2 R_s = rac{V^2}{R_s^2 + X_s^2} R_s$$

$$Q_s = |I_s|^2 X_s = rac{V^2}{R_s^2 + X_s^2} X_s$$

For the parallel circuit, since the real and reactive powers of both circuits will be the same, we have

$$R_{p} = \frac{V^{2}}{P} = \frac{R_{s}^{2} + X_{s}^{2}}{R_{s}}$$
$$X_{p} = \frac{V^{2}}{Q} = \frac{R_{s}^{2} + X_{s}^{2}}{X_{s}}$$

(2-7) As calculated in Example 2-2,

$$Z_{in} = 6.775 \angle 29.03^{\circ}$$
$$= 5.924 + j3.289$$

Then, we will have,

$$R_P = \frac{R_S^2 + X_S^2}{R_S} = \frac{5.924^2 + 3.289^2}{5.924} = 7.75\Omega$$
$$X_P = \frac{R_S^2 + X_S^2}{X_S} = \frac{5.924^2 + 3.289^2}{3.289} = 13.96\Omega$$

(2-8) In Example 2-2

$$\overline{I_1} = 17.172 \angle -29.03^{\circ} A$$

$$\overline{I_{Rp}} = \frac{jX_P}{R_P + jX_P} \overline{I_1} = \frac{j13.96}{7.75 + j13.96} \times 17.172 \angle -29.03 = 15.483 \angle 0^{\circ} A$$

$$\overline{I_{Xp}} = \frac{R_P}{R_P + jX_P} \overline{I_1} = \frac{7.75}{7.75 + j13.96} \times 17.172 \angle -29.03 = 8.595 \angle 89.99^{\circ} A$$

$$P = I_{Rp}^2 R_P = 15.483^2 \times 7.75 = 1857.86W$$

$$Q = I_{Xp}^2 X_P = 8.595^2 \times 13.96 = 1031.28VAR$$

Thus, P and Q are the same as calculated in Ex. 2-3. Hence the R_P and X_P calculations are correct.

(2-9) Referring to Figure 1,

$$\overline{V} = V \angle 0^{\circ}, I_{net} = I \angle \phi,$$

$$Z_{net} = \frac{\overline{V}}{I} = |Z| \angle -\phi, cos\phi = 0.9$$

$$sin(\phi) = \sqrt{1 - cos^2 \phi} = 0.436$$

$$\therefore Q_{net} = |P_L tan\phi| = 900.29 VAR$$

$$Q_C = Q_{net} + Q_L = 1931.59 VAR$$

$$X_C = \frac{V^2}{Q_C}$$

$$\therefore \frac{1}{(2\pi f)C} = \frac{V^2}{Q_C}$$

$$C = \frac{Q_C}{(2\pi f)V^2} = 356 \mu F$$

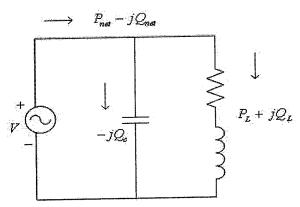


Figure 1:

(2-10) As shown in Figure 2,

$$Q_{net} = 5000 \, tan[cos^{-1}(0.95)] = 1643.42 VAR$$
 $Q_{net} = Q_L - Q_C$
 $Q_C = Q_L - Q_{net} = 2106.58 VAR$
 $X_C = \frac{V^2}{Q_C} = 6.835\Omega$
 $\therefore C = \frac{1}{(2\pi f)X_C} = 388 \mu F$

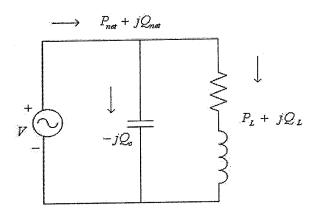


Figure 2:

(2-11) Note that

$$\overline{V}_a = 100 \angle 30^{\circ} V$$

Thus,

$$\begin{split} \overline{V}_b &= 100 \angle -90^\circ V, \\ \overline{V}_c &= 100 \angle 150^\circ V \\ v_a(t) &= \sqrt{2} \times 100 cos(\omega t + 30) = 141.4 cos(\omega t + 30) \\ v_b(t) &= \sqrt{2} \times 100 cos(\omega t + 30 - 120) = 141.4 cos(\omega t - 90) \\ v_c(t) &= \sqrt{2} \times 100 cos(\omega t + 30 + 120) = 141.4 cos(\omega t + 150) \\ \overline{V}_{ab} &= \overline{V}_a - \overline{V}_b = 173.2 \angle 60^\circ \\ v_{ab}(t) &= \sqrt{2} \times 173.2 cos(\omega t + 60) = 244.9 cos(\omega t + 60) \end{split}$$

(2-12) Given a Y-connected inductive load,

$$3V_a I_a cos \phi = 10000$$

$$\therefore I_a = \frac{10000}{3 \times 120 \times 0.9}$$

$$= 30.86A$$

Assuming $V_a = 120 \angle 0^\circ$

$$Z = |Z| \angle cos^{-1}(0.9) = |Z| \angle 25.84\Omega$$
$$\therefore |Z| = \frac{V_a}{I_a} = 3.89\Omega$$

$$\overline{I}_a = \frac{\overline{V}_a}{Z} = \frac{120\angle - 25.84^{\circ}}{3.89} = 30.86\angle - 25.84^{\circ}A$$

$$\overline{I}_b = 3.86\angle - 25.84 - 120^{\circ}3.89 = 30.86\angle - 145.84^{\circ}A$$

$$\overline{I}_c = 3.86\angle - 25.84 + 120^{\circ}3.89 = 30.86\angle 94.16^{\circ}A$$

(2-13) Same as above , only magnitude of load impedance changes

$$\therefore |Z| = 3.89 \times 3 = 11.67\Omega$$

(2-14)

$$\overline{V_{aA}} = (Z_{self} - Z_{mutual})\overline{I_a} = (0.3 + j1) \times 10 \angle -30^{\circ}$$
$$= 10.44 \angle 43.3^{\circ} V$$

$$\overline{V_{aA}} = \overline{V_a} - \overline{V_A}$$

$$= 1000 \angle 0^\circ - \overline{V_A}$$

$$\therefore \overline{V_A} = 1000 \angle 0^\circ - 10.44 \angle 43.3^\circ$$

$$= 992.43 \angle - 0.413^\circ V$$

(2-15) In a balanced circuit,

$$\begin{split} \overline{V_a} + \overline{V_b} + \overline{V_c} &= 0 \\ \overline{V_b} + \overline{V_c} &= -\overline{V_a} \\ \overline{I_a} &= \frac{\overline{V_a}}{-jX_{c2}} + \frac{\overline{V_a} - \overline{V_b}}{-jX_{c1}} + \frac{\overline{V_a} - \overline{V_c}}{-jX_{c1}} \\ &= \frac{3\overline{V_a}}{-jX_{c1}} + \frac{\overline{V_a}}{-jX_{c2}} \\ &= \frac{\overline{V_a}}{Z_{eq}} \text{ where } Z_{eq} = \frac{1}{\frac{3}{-jX_{c1}} + \frac{1}{-jX_{c2}}} \end{split}$$

(2-16)

$$\begin{split} P_R &= \frac{V_S V_R}{X} sin\delta \\ sin\delta &= \frac{1000 \times 1.5}{100 \times 95} = 0.158 \\ \therefore \delta &= 9.091^{\circ} \\ \overline{I} &= \frac{\overline{V_S} - \overline{V_R}}{jX} = \frac{100 \angle \delta - 95 \angle 0}{j1.5} \\ &= \frac{100 \angle 9.091 - 95 \angle 0}{1.5 \angle 90} \\ &= 10.83 \angle - 13.39^{\circ} \end{split}$$

From Eq 2-46

$$Q_R = \frac{V_S V_R}{X} cos\delta - \frac{V_R^2}{X} = 237.11 VAR$$

As a check

$$P_R + jQ_R = \overline{V_R} \overline{I}^* = 1028.85 \angle 13.39^\circ$$

 $\therefore Q_R = 238.3 VAR$

(2-17)

$$\begin{aligned} P &= |V_S||I|cos\phi = 100|I|cos\phi \\ \overline{V_R} &= \overline{V_S} - jX\overline{I} \end{aligned}$$

Solving the above two equations we get an expression for $\frac{|V_S|}{|V_R|}$ and a corresponding plot in Matlab as shown above

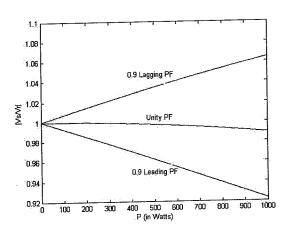


Figure 3:

(2-18)
$$V_{LL,base}$$
 = 208 V, $V_{ph,base}$ = $\frac{208}{\sqrt{3}}$ = 120 V, $P_{ph,base}$ = 1200 W
$$I_{base} = \frac{P_{ph,base}}{V_{ph,base}} = 10A$$

$$Z_{base} = \frac{V_{ph,base}}{I_{base}} = 12\Omega$$

With these base values, we calculate the per unit values as,

$$\overline{V_{ph}} = \frac{120\angle 0}{120} = 1.0\angle 0^{\circ} pu$$

$$\overline{I_L} = \frac{12\angle - 36.87^{\circ}}{10} = 1.2\angle - 36.87^{\circ} pu$$

$$\overline{Z_L} = \frac{10\angle 36.87^{\circ}}{12} = 0.83\angle 36.87^{\circ} pu$$

$$P_L = \frac{1152}{1200} = 0.96pu$$

$$Q_L = \frac{864}{1200} = 0.72pu$$

(2-19)
$$V_{LL,base}$$
 = 240 V, $V_{ph,base}$ = 138.57 V, $P_{ph,base}$ = 1800 W
$$I_{base} = \frac{P_{ph,base}}{V_{ph,base}} = 12.99 A$$

$$Z_{base} = \frac{V_{ph,base}}{I_{hase}} = 10.67 \Omega$$

With these base values, we calculate the per unit values as,

$$\overline{V_{ph}} = 0.867 \angle 0^{\circ} pu$$

$$\overline{I_L} = \frac{138.57}{|Z_L|} \angle - 36.87^{\circ} = 13.86 \angle - 36.87^{\circ}$$

$$\overline{I} = \frac{13.86 \angle 36.87^{\circ}}{12.99} = 1.067 \angle - 36.87^{\circ} pu$$

$$\overline{Z_L} = \frac{10 \angle 36.87^{\circ}}{10.67} = 0.937 \angle 36.87^{\circ} pu$$

$$P_L = \frac{V^2}{|Z_L|} \cos \phi = 1536.13W = 0.853pu$$

$$Q_L = \frac{V^2}{|Z_L|} \sin \phi = 1152.1VAR = 0.64pu$$

(2-20)
$$V_{LL,base}=240$$
 V, $V_{ph,base}=138.57$ V, $P_{ph,base}=1200$ W
$$I_{base}=\frac{P_{ph,base}}{V_{ph,base}}=8.66A$$
 $Z_{base}=\frac{V_{ph,base}}{I_{base}}=16\Omega$

With these base values, we calculate the per unit values as,

$$\overline{V_{ph}} = 0.867 \angle 0^{\circ} pu$$

$$\overline{I_L} = \frac{138.57}{|Z_L|} \angle - 36.87^{\circ} = 13.86 \angle - 36.87^{\circ}$$

$$\overline{I} = \frac{13.86 \angle 36.87^{\circ}}{8.66} = 1.6 \angle - 36.87^{\circ} pu$$

$$\overline{Z_L} = \frac{10 \angle 36.87^{\circ}}{16} = 0.625 \angle 36.87^{\circ} pu$$

$$P_L = \frac{V^2}{|Z_L|} \cos \phi = 1536.13W = 1.28pu$$

$$Q_L = \frac{V^2}{|Z_L|} \sin \phi = 1152.1VAR = 0.96pu$$

(2-21)(a)

$$l_m = 2\pi r_m$$

$$H_{ID} = \frac{Ni}{l_m} = 477.71A/m$$

(b)

$$l_m = \pi OD$$

$$H_{OD} = \frac{Ni}{l_m} = 433.53A/m$$

(c)

$$Error_{ID} = \frac{H_{ID} - H_m}{H_m} \times 100 = 5.107\%$$
 $Error_{OD} = \frac{H_{OD} - H_m}{H_m} \times 100 = 4.614\%$

(2-22)

$$\mathcal{R}_{m} = \frac{l_{m}}{\mu_{0}\mu_{r}A_{m}}$$

$$= \frac{0.165}{4\pi \times 10^{-7} \times 2000 \times \frac{\pi}{4} \left(\frac{OD-ID}{2}\right)^{2}}$$

$$= 1.337 \times 10^{7} A/Wb$$

(2-23) Using the \mathcal{R}_m from Prob 2-22,

$$L = \frac{N^2}{\mathcal{R}_m}$$

$$\therefore N = \sqrt{L\mathcal{R}_m} = 19$$

$$B_m = \mu_0 \mu_r \frac{NI}{l_m}$$

$$\therefore \mu_r = \frac{B_m l_m}{\mu_0 NI} = \frac{1.3 \times 0.165}{4\pi \times 10^{-7} \times 19 \times 3} = 2994.63$$

(2-24)

$$\mathcal{R} = \mathcal{R}_m + \mathcal{R}_{gap} \approx \mathcal{R}_{gap} = rac{l_{gap}}{\mu_{gap} A_{gap}} ext{ [Using } A_{gap} ext{ from Prob 2-22]}$$

$$= rac{l_{gap}}{4\pi \times 10^{-7} \times 4.91 \times 10^{-6}} = 16.2 \times 10^{10} l_{gap}$$

$$\lambda = N\varphi = LI$$

$$\therefore NB_{max}A_m = LI$$

$$N = \frac{LI}{B_{max}A_m}$$

$$= \frac{25 \times 10^{-6} \times 3}{1.3 \times 4.91 \times 10^{-6}} = 12$$

$$\mathcal{R} = \frac{N^2}{L}$$

$$\therefore 16.2 \times 10^{10} l_{gap} = \frac{N^2}{25 \times 10^{-6}}$$

$$l_{gap} = \frac{N^2}{25 \times 10^{-6} \times 16.2 \times 10^{10}} = 0.00356 cm$$